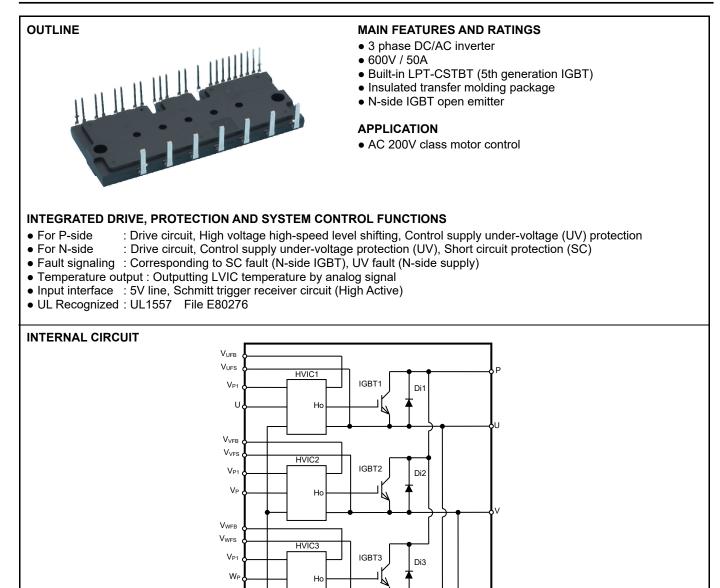


< DIPIPM > PS21A79 TRANSFER MOLDING TYPE

INSULATED TYPE



VPC

 V_{N1}

UN

VΝ

WN

Fo Vot Vnc LVIC

Uout

Vou

Wout

CEO

IGBT4

IGBT5

IGBT6

Di4

Di5

Di6

CIN Vsc

NU

Nν

NW

MAXIMUM RATINGS (T_j = 25°C, unless otherwise noted)

INVERTER PART

		0	D ^r	1.1.14
Symbol	Parameter	Condition	Ratings	Unit
V _{cc}	Supply voltage	Applied between P-NU,NV,NW	450	V
V _{CC(surge)}	Supply voltage (surge)	Applied between P-NU,NV,NW	500	V
V _{CES}	Collector-emitter voltage		600	V
±l _c	Each IGBT collector current	T _c = 25°C	50	А
±I _{CP}	Each IGBT collector current (peak)	T_c = 25°C, up to 1ms	100	Α
Pc	Collector dissipation	T _c = 25°C, per 1 chip	142	W
Tj	Junction temperature		-20~+150	°C

CONTROL (PROTECTION) PART

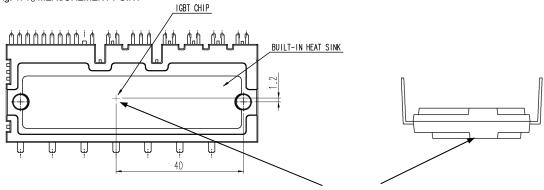
Symbol	Parameter	Condition	Ratings	Unit
VD	Control supply voltage	Applied between V_{P1} - V_{PC} , V_{N1} - V_{NC}	20	V
V _{DB}	Control supply voltage	Applied between V _{UFB} -V _{UFS} , V _{VFB} -V _{VFS} , V _{WFB} -V _{WFS}	20	V
V _{IN}	Input voltage	Applied between U _P , V _P , W _P -V _{PC} , U _N , V _N , W _N -V _{NC}	-0.5~V _D +0.5	V
V _{FO}	Fault output supply voltage	Applied between Fo-VNC	-0.5~V _D +0.5	V
I _{FO}	Fault output current	Sink current at F _o terminal	1	mA
Vsc	Current sensing input voltage	Applied between CIN-V _{NC}	-0.5~V _D +0.5	V

TOTAL SYSTEM

Symbol	Parameter	Condition	Ratings	Unit
V _{CC(PROT)}	Self protection supply voltage limit (Short circuit protection capability)	V_D = 13.5~16.5V, Inverter Part T _i = 125°C, non-repetitive, up to 2µs	400	V
Tc	Module case operation temperature	(Note 1)	-20~+100	°C
T _{stg}	Storage temperature		-40~+125	°C
V _{iso}	Isolation voltage	60Hz, Sinusoidal, AC 1min, between connected all pins and heat sink plate	2500	V _{rms}

Note 1: Tc measurement point is described in Fig.1.

Fig. 1: T_C MEASUREMENT POINT



Measurement point for Tc

THERMAL RESISTANCE

Symbol	Parameter	Condition	Limits			Unit
Symbol	Farameter	Condition		Тур.	Max.	Unit
R _{th(j-c)Q}	Junction to case thermal	Inverter IGBT part (per 1/6 module)	-	-	0.88	K/W
R _{th(j-c)F}	resistance (Note 2)	Inverter FWDi part (per 1/6 module)	-	-	1.78	K/W

Note 2: Grease with good thermal conductivity and long-term endurance should be applied evenly with about +100µm~+200µm on the contacting surface of DIPIPM and heat sink. The contacting thermal resistance between DIPIPM case and heat sink Rth(c-f) is determined by the thickness and the thermal conductivity of the applied grease. For reference, Rth(c-f) is about 0.2K/W (per 1/6 module, grease thickness: 20µm, thermal conductivity: 1.0W/m•k).

ELECTRICAL CHARACTERISTICS (T_j = 25°C, unless otherwise noted) **INVERTER PART**

Sympol	Parameter	Condition			Limits		
Symbol	Parameter	Condit	1011	Min.	Тур.	Max.	Unit
V	Collector-emitter saturation	V _D =V _{DB} = 15V, V _{IN} = 5V, I _C = 50A	T _j = 25°C	-	1.55	2.05	v
$V_{\text{CE(sat)}}$	voltage	$v_{\rm D} - v_{\rm DB} = 15v, v_{\rm IN} - 5v, I_{\rm C} - 50A$	T _j = 125°C	-	1.65	2.10	v
V _{EC}	FWDi forward voltage	V _{IN} = 0V, -I _C = 50A		-	1.70	2.20	V
t _{on}				1.80	2.40	3.60	μs
t _{C(on)}		V _{CC} = 300V, V _D = V _{DB} = 15V		-	0.40	0.60	μs
t _{off}	Switching times			-	3.00	4.20	μs
t _{C(off)}				-	0.60	1.20	μs
t _{rr}				-	0.30	-	μs
	Collector-emitter cut-off	<u>)</u>	T _j = 25°C	-	-	1	
ICES	current	V _{CE} =V _{CES}	T _j = 125°C	-	-	10	mA

CONTROL (PROTECTION) PART

Symbol	Parameter		Condition			Limits		Unit
Symbol	Falameter		Condition		Min.	Тур.	Max.	Onit
1		Tatal of $V_{\rm D}$ = 15V, $V_{\rm IN}$ = 0V			-	-	5.50	
ID		Total of V_{P1} - V_{PC} , V_{N1} - V_{NC}	V_D =15V, V_{IN} =5V		-	-	5.50	mA
1	Circuit current	Each part of V _{UFB} -V _{UFS} ,	$V_D = V_{DB} = 15V, V_{IN} = 0V$		-	-	0.55	mA
I _{DB}		VVFB-VVFS, VWFB-VWFS	$V_D = V_{DB} = 15V, V_{IN} = 5V$		-	-	0.55]
I _{SC}	Short circuit trip level		-20°C≤Tj≤125°C, Rs=40.2Ω (±1%), Not connecting outer shunt resistors to (Note 3) NU.NV.NW terminals		85	-	-	A
UV _{DBt}	P-side Control supply	T (105%)	Trip level		10.0	-	12.0	V
UV_{DBr}	 under-voltage protection(UV) 	T _j ≤125°C	Reset level		10.5	-	12.5	V
UV_{Dt}	N-side Control supply	T <105%0	Trip level		10.3	-	12.5	V
UV _{Dr}	under-voltage protection(UV)	T _j ≤125°C	Reset level		10.8	-	13.0	V
V _{FOH}		V _{SC} = 0V, F _O terminal pulled	d up to 5V by 10kΩ		4.9	-	-	V
V _{FOL}	Fault output voltage	V_{SC} = 1V, I_{FO} = 1mA			-	-	0.95	V
t _{FO}	Fault output pulse width	C _{FO} =22nF		(Note 4)	1.6	2.4	-	ms
I _{IN}	Input current	$V_{IN} = 5V$			0.70	1.00	1.50	mA
V _{th(on)}	ON threshold voltage	Applied between LL V W	Applied between U _P , V _P , W _P , U _N , V _N , W _N -V _{NC}		2.1	2.3	2.6	v
V _{th(off)}	OFF threshold voltage	Applied between OP, VP, VV	P, UN, VN, VVN-VNC		0.8	1.4	2.1	v
V _{OT}	Temperature output	LVIC temperature = 85°C		(Note 5)	3.57	3.63	3.69	V

Note 3: Short circuit protection detects sense current divided from main current at N-side IGBT and works for N-side IGBT only. In the case that outer shunt resistor is inserted into main current path, protection current level Isc changes. For details, please refer the application note for this DIPIPM.

4: Fault signal is output when short circuit or N-side control supply under-voltage protection works. The fault output pulse-width t_{FO} depends on the capacitance of C_{FO} . (C_{FO} (typ.) = $t_{FO} \times (9.1 \times 10^6)$ [F]) 5: DIPIPM doesn't shutdown IGBTs and output fault signal automatically when temperature rises excessively. When temperature exceeds the protective level that user defined, controller (MCU) should stop the DIPIPM immediately. This output might reach 5V when temperature rises excessively, so in the case of using low voltage controller like 3.3V MCU, it is recommended to insert a clamp Di between controller supply (e.g. 3.3V) and V_{OT} output for overvoltage protection. Temperature of LVIC vs. V_{OT} output characteristics and V_{OT} output circuit are described in Fig.2 and Fig.3.

Fig. 2 Temperature of LVIC vs. Vot Output Characteristics

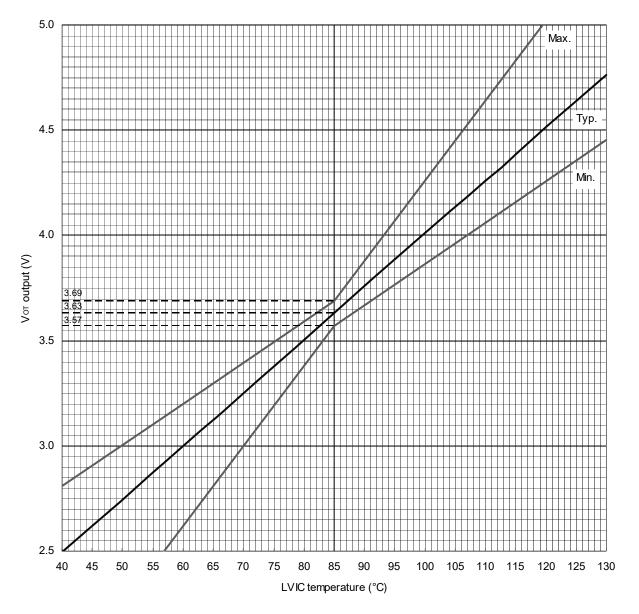
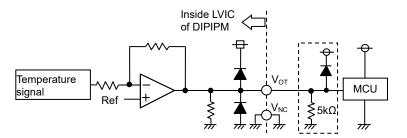


Fig.3 Vot Output Circuit

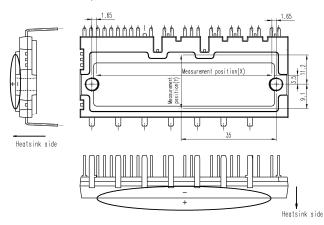


- (1) It is recommended to insert $5k\Omega$ or more (5.1k Ω is recommended.) pull down resistor for getting linear output characteristics at lower temperature than room temperature. When the pull down resistor is inserted between V_{OT} and V_{NC}(control GND), the extra current calculated by V_{OT} output voltage / pull down resistance flows as LVIC circuit current continuously. In the case of only using V_{OT} for detecting higher temperature than room temperature, it isn't necessary to insert the pull down resistor.
- (2) This output might reach 5V when temperature rises excessively, so in the case of using low voltage controller like 3.3V MCU, it is recommended to insert a clamp Di between controller supply (e.g. 3.3V) and V_{OT} output for overvoltage protection.
 (3) If V_{OT} output function is unnecessary, leave this output terminal open (no connection).
- Please refer the application note about the usage of $V_{\mbox{\scriptsize OT}}$ too.

MECHANICAL CHARACTERISTICS AND RATINGS

Deremeter	Condition			1.1				
Parameter		nation	Min.	Тур.	Max.	Unit		
Mounting torque	Mounting screw : M4	Recommended 1.18N·m	0.98	-	1.47	N∙m		
Terminal pulling strength	Load 19.6N	JEITA-ED-4701	10	-	-	s		
Terminal bending strength	Load 9.8N, 90deg. bend	JEITA-ED-4701	2	-	-	times		
Weight			-	46	-	g		
Heat-sink flatness		-50	-	100	μm			

Note 6: Measurement point of heat-sink flatness



RECOMMENDED OPERATION CONDITIONS

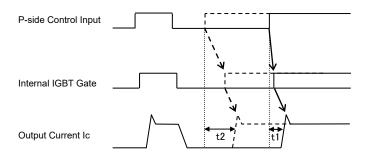
Sympol	Parameter	Condition			Limits		Unit	
Symbol	Parameter	Condition		Min.	Тур.	Max.	Onit	
V _{cc}	Supply voltage	Applied between P-NU, NV, NW		0	300	400	V	
VD	Control supply voltage	Applied between V _{P1} -V _{PC} , V _{N1} -V _{NC}		13.5	15.0	16.5	V	
V _{DB}	Control supply voltage	Applied between V _{UFB} -V _{UFS} , V _{VFB} -V _{VFS} , V _V	_{/FB} -V _{WFS}	13.0	15.0	18.5	V	
$\Delta V_D, \Delta V_{DB}$	Control supply variation			-1	-	+1	V/µs	
t _{dead}	Arm shoot-through blocking time	For each input signal		2.2	-	-	μs	
f _{PWM}	PWM input frequency	T _C ≤ 100°C, T _j ≤ 125°C		-	-	20	kHz	
	Allowable r m a current	wable r.m.s. current $ \begin{array}{l} V_{\rm CC}=300V, V_{\rm D}=15V, {\rm P.F}=0.8, \\ {\rm Sinusoidal} {\rm PWM} \\ {\rm T_C}\leq 100^\circ {\rm C}, {\rm T_j}\leq 125^\circ {\rm C} \qquad ({\rm Note} \ 7) \end{array} \begin{array}{l} f_{\rm PWM}=5 {\rm kHz} \\ f_{\rm PWM}=15 {\rm kHz} \end{array} $	f _{PWM} = 5kHz	-	-	23.6	Arms	
l _o	Allowable I.III.S. current		f _{PWM} = 15kHz	-	-	13.8	Anns	
PWIN(on)			(Note 8)	1.1	-	-		
PWIN(off)	Minimum input pulse width	200≤ V _{CC} ≤ 350V, 13.5≤ V _D ≤ 16.5V, 13.0≤ V _{DB} ≤ 18.5V, -20°C ≤ T _C ≤ 100°C,	I _C ≤50A	3.0	-	-	μs	
		N line wiring inductance less than 10nH (Note 9) 50A <i<sub>C≤85A</i<sub>		5.0	-	-		
V _{NC}	V _{NC} variation	Between V_{NC} -NU, NV, NW (including surg	je)	-5.0	-	+5.0	V	
Tj	Junction temperature			-20	-	+125	°C	

Note

7: The allowable r.m.s. current value depends on the actual application conditions.
8: DIPIPM might not make response to the input on signal with pulse width less than PWIN (on).
9: IPM might make no response or delayed response (P-side IGBT only) for the input signal with off pulse width less than PWIN(off).

Please refer Fig. 4 about delayed response.

Fig. 4 About Delayed Response Against Shorter Input Off Signal Than PWIN(off) (P-side only)



Solid line

Broken line

⋯Off pulse width ≥ PWIN(off); Turn on time t1 (Normal delay) ⋯Off pulse width < PWIN(off);</p>

Turn on time t2 (Longer delay in some cases)

Fig. 5 Timing Charts of DIPIPM Protective Functions

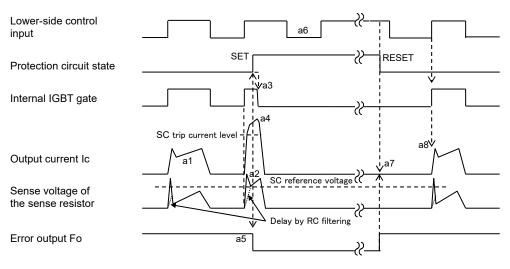
- [A] Short-Circuit Protection (N-side only with the external sense resistor and RC filter)
- a1. Normal operation: IGBT ON and outputs current.
- a2. Short circuit current detection (SC trigger)

(It is recommended to set RC time constant 1.5~2.0µs so that IGBT shut down within 2.0µs when SC occurs.)

- a3. All N-side IGBT's gates are hard interrupted.
- a4. All N-side IGBTs turn OFF.
- a5. F_{O} outputs with a fixed pulse width determined by the external capacitor $C_{\text{FO}}.$

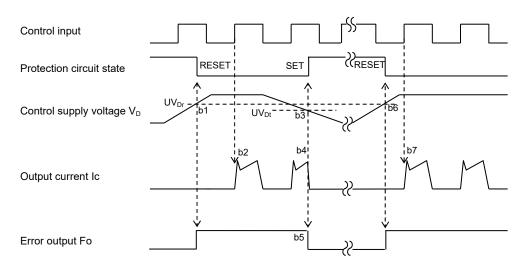
a6. Input = "L": IGBT OFF

- a7. Fo finishes output, but IGBTs don't turn on until inputting next ON signal (L \rightarrow H).
- (IGBT of each phase can return to normal state by inputting ON signal to each phase.)
- a8. Normal operation: IGBT ON and outputs current.



[B] Under-Voltage Protection (N-side, UV_D)

- b1. Control supply voltage V_D exceeds under voltage reset level (UV_{Dr}), but IGBT turns ON by next ON signal (L \rightarrow H). (IGBT of each phase can return to normal state by inputting ON signal to each phase.)
- b2. Normal operation: IGBT ON and outputs current.
- b3. V_D level drops to under voltage trip level. (UV_{Dt}).
- b4. All N-side IGBTs turn OFF in spite of control input condition.
- b5. Fo outputs for the period determined by the capacitance C_{FO_1} but output is extended during V_D keeps below UV_{Dr} .
- b6. V_D level reaches UV_{Dr}.
- b7. Normal operation: IGBT ON and outputs current by next ON signal (L \rightarrow H).



[C] Under-Voltage Protection (P-side, UV_{DB})

- c1. Control supply voltage V_{DB} rises. After the voltage reaches under voltage reset level UV_{DBr}, IGBT turns on by next ON signal (L→H).
- c2. Normal operation: IGBT ON and outputs current.
- c3. V_{DB} level drops to under voltage trip level (UV_{DBt}).
- c4. IGBT of corresponding phase only turns OFF in spite of control input signal level, but there is no Fo signal output.
- c5. V_{DB} level reaches UV_{DBr} .
- c6. Normal operation: IGBT ON and outputs current by next ON signal (L \rightarrow H).

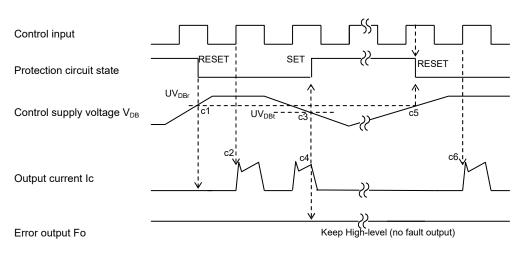
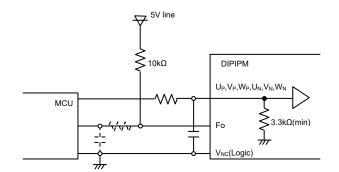


Fig. 6 MCU I/O Interface Circuit

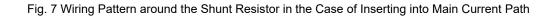


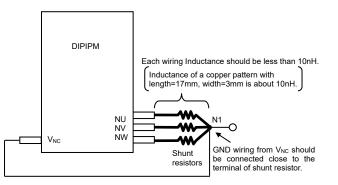
Note)

Design for input RC filter depends on the PWM control scheme used in the application and the wiring impedance of the printed circuit board.

The DIPIPM input signal interface integrates a min. $3.3k\Omega$ pull-down resistor. Therefore, when using RC filter, be careful to satisfy turn-on threshold voltage requirement.

 $\dot{F}o$ output is open drain type. It should be pulled up to the positive side of 5V or 15V power supply with the resistor that limits Fo sink current I_{Fo} under 1mA. In the case of pulling up to 5V supply, over $5.1 k\Omega$ is needed. (10k Ω is recommended.)

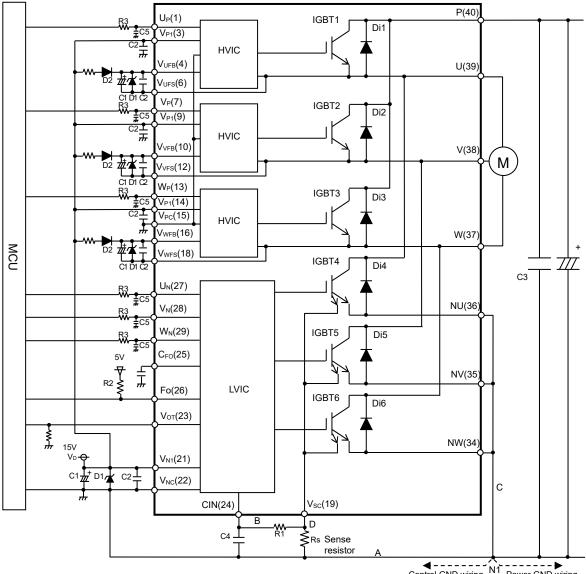




Low inductance shunt resistor like surface mounted (SMD) type is recommended. Protection current level I_{SC} changes by inserting shunt resistor.

< DIPIPM > **PS21A79** TRANSFER MOLDING TYPE **INSULATED TYPE**

Fig. 8 Example of Application Circuit



Note

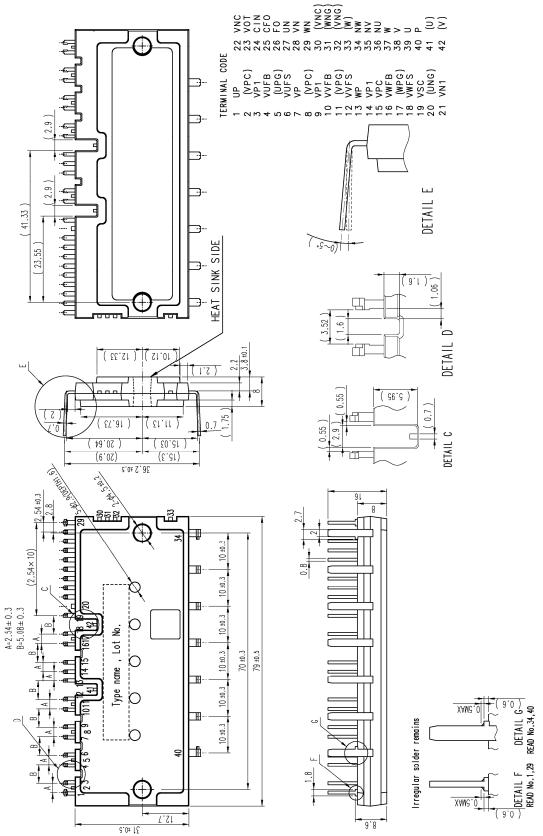
- 1 :If control GND and power GND are patterned by common wiring, it may cause malfunction by fluctuation of power GND level. It is recommended to connect control GND and power GND at only a N1 point at which NU, NV, NW are connected to power GND line.
- 2 :It is recommended to insert a Zener diode D1 (24V/1W) between each pair of control supply terminals to prevent surge destruction.
- 3:To prevent surge destruction, the wiring between the smoothing capacitor and the P, N1 terminals should be as short as possible. Generally inserting a 0.1µ~0.22µF snubber capacitor C3 between the P-N1 terminals is recommended.
- 4 :R1, C4 of RC filter for preventing protection circuit malfunction is recommended to select tight tolerance, temp-compensated type. The time constant R1C4 should be set so that SC current is shut down within 2µs. (1.5µs~2µs is general value.) SC interrupting time might vary with the wiring pattern, so the enough evaluation on the real system is recommended. If R1 is too small, it may leads to delay of protection. So R1 should be min. 10 times larger resistance than Rs. (100 times is recommended.)
- 5 To prevent erroneous operation, the wiring of A, B, C should be as short as possible.
- 6 :For sense resistor, the variation within 1% (including temperature characteristics), low inductance type is recommended. And the over 1/8W is recommended, but it is necessary to evaluate in your real system finally.
- 7 :To prevent erroneous SC protection, the wiring from V_{SC} terminal to CIN filter should be divided at the point D that is close to the terminal of sense resistor. And the wiring should be patterned as short as possible.
- 8 :All capacitors should be mounted as close to the terminals of the DIPIPM as possible. (C1: good temperature, frequency characteristic electrolytic type, and C2: 0.22µ~2.0µF, good temperature, frequency and DC bias characteristic ceramic type are recommended.)
- 9 :Input drive is High-active type. There is a min. 3.3kΩ pull-down resistor in the input circuit of IC. To prevent malfunction, the wiring of each input should be as short as possible. And it is strongly recommended to insert RC filter (e.g. R3=100Ω and C5=1000pF) and confirm the input signal level to meet the turn-on and turn-off threshold voltage. Thanks to HVIC inside the module, direct coupling to MCU without any opto-coupler or transformer isolation is possible.
- 10 :Fo output is open drain type. It should be pulled up to MCU or control power supply (e.g. 5V,15V) by a resistor that makes IFo up to 1mA. (IFO is estimated roughly by the formula of control power supply voltage divided by pull-up resistance. In the case of pulled up to 5V, 10kΩ (5kΩ or more) is recommended.)
- 11 :Error signal output width (t_{Fo}) can be set by the capacitor connected to C_{FO} terminal. $C_{FO}(typ.) = t_{Fo} x (9.1 \times 10^{-6}) (F)$
- 12 :High voltage (VRRM =600V or more) and fast recovery diode (trr=less than 100ns or less) should be used for D2 in the bootstrap circuit.
- 13 :If high frequency noise superimposed to the control supply line, IC malfunction might happen and cause erroneous operation. To avoid such problem, voltage ripple of control supply line should meet dV/dt ≤+/-1V/µs, Vripple≤2Vp-p.
- 14 :For DIPIPM, it isn't recommended to drive same load by parallel connection with other phase IGBT or other DIPIPM.

Power GND wiring Control GND wiring

< DIPIPM > PS21A79 TRANSFER MOLDING TYPE INSULATED TYPE

Fig. 9 Package Outlines

Dimensions in mm



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